Disclosing the Complexity of Nonlinear Ship Rolling and Duffing Oscillators by a Signum Function

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Abstract: In this paper we study the nonlinear dynamical system $\dot{x} = f(x, t)$ from a newly developed theory, viewing the time-varying function of $\text{sign}(|f|^2 |x|^2 - 2(f \cdot x)^2) = -\text{sign}(\cos 2\theta)$ as a key factor, where $\theta$ is the intersection angle between $x$ and $f$. It together with $\text{sign}(\cos \theta)$ can reveal the complexity of nonlinear Duffing oscillator and a quadratic ship rolling oscillator. The barcode is formed by plotting $\text{sign}(|f|^2 |x|^2 - 2(f \cdot x)^2)$ with respect to time. We analyze the barcode to point out the bifurcation of subharmonic motions and the range of chaos in the parameter space. The bifurcation diagram obtained by plotting the percentage of the first set of dis-connectivity $A1^{-}:=\{\text{sign}(\cos \theta) = +1 \text{ and } \text{sign}(\cos 2\theta) = +1\}$ with respect to the amplitude of harmonic loading leads to a finer structure of a devil staircase for the ship rolling oscillator, as well as a cascade of subharmonic motions to chaos for the Duffing oscillator.

Keywords: Nonlinear dynamical system, $SO_2(n, 1)$ Lie-group scheme, Signum function, Barcode, Duffing oscillator, Ship rolling equation, Jumping in Barcode, First set of dis-connectivity

1 Introduction

A cubic nonlinear oscillator described by the Duffing equation has the following form:

$$\ddot{x}(t) + \gamma \dot{x}(t) + \alpha x(t) + \beta x^3(t) = F_0 \cos(\omega t),$$

(1)

where $x$ is the displacement, $\gamma$ is a damping constant, $F_0$ and $\omega$ are respectively the amplitude and excitation frequency of a harmonic loading, and $\sqrt{\alpha}$ with $\alpha > 0$ is the frequency of natural oscillation of the corresponding linear system.

The Duffing equation has been a major subject of intensive study over the last few decades as one of innocent examples of nonlinear dynamical system exhibiting

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chaotic behavior, and which lends a typical chaotic system with a period doubling route to chaos. The study of nonlinear oscillators is of great importance not only in all areas of physics but also in engineering and other disciplines, since most phenomena in our world are nonlinear and are described by nonlinear equations. Recently, a considerable attention has been directed towards the semi-analytical solutions for nonlinear oscillators. There are many computational methods that have been developed for solving the periodic and sub/super-harmonic solutions of the nonlinear oscillators, for example, the harmonic balance method [Donescu1, Virgin and Wu (1996); Wu, Sun and Lim (2006), Liu, Thomas, Dowell, Attar and Hall (2006)], the variational iteration method [He (1999); Ozis and Yildirim (2007)], the homotopy perturbation method [He (2000); Shou (2009)], the parameter-expanding method [Koroglu and Ozis (2011)], the exp-function method [He and Abdou (2007)], and differential transform method [Chu and Lo (2011)]. Recently, Dai, Schnoor and Atluri (2012) have applied a simple collocation method to reveal the complex subharmonic behavior of the Duffing oscillator, and Liu (2012) proposed a Lie-group adaptive method to solve the optimal control problem of nonlinear Duffing oscillator.

The chaotic behavior of ship rolling motion in beam sea has been studied by Thompson (1997), of which a typical equation to explore the instability of ship capsize is the following quadratic nonlinear oscillator:

$$\ddot{x}(t) + \gamma \dot{x}(t) + x(t) - x^2(t) = F_0 \sin(\omega t).$$

In the studies of ship motion the analysis of large amplitude nonlinear rolling motion is important to understand the capsize dynamics.

The analyses of Eqs. (1) and (2) have been performed to obtain approximate solutions by using various methods, like multiple scale method, perturbation method, harmonic balance method, the Bogouibov Mitropolsky asymptotic method. The numeric safe basins, the Melnikov method, the Lyapunov exponents and the Lyapunov direct method were used to determine the conditions of stability and the occurrence of chaotic motion. The analytic methods used to predict chaos [Szemplinska - Stupnicka (1995); Litak and Borowiec (2006)] are also developed by using an observation of period doubling, and Melnikov’s method based on the phenomenon of homoclinic intersection. But there are still many problems that the existent methods cannot exhaust the complicated behavior of nonlinear dynamical systems.
2 A nonlinear augmented system formulation

To facilitate the formulation we write the above equations as a system of first-order ordinary differential equations (ODEs):

\[ \dot{x} = f(x, t), \quad t \in \mathbb{R}, \quad x \in \mathbb{R}^n, \]  

which is an \( n \)-dimensional ODEs system.

As that done by Liu (2001), for Eq. (3) we can define a unit orientation vector:

\[ n := \frac{x}{\|x\|}, \]  

where \( \|x\| := \sqrt{x \cdot x} > 0 \) is the Euclidean norm of \( x \), and the dot between two vectors, say \( x \cdot y \), denotes the inner product of \( x \) and \( y \).

First, upon using Eqs. (3) and (4) the length \( \|x\| \) is governed by

\[ \frac{d}{dt} \|x\| = \frac{\dot{x} \cdot x}{\sqrt{x \cdot x}} = f(x, t) \cdot n. \]
Then, using Eqs. (3)-(5) we can derive

\[
\dot{n} = \frac{f(x,t)}{\|x\|} - \left( \frac{f(x,t)}{\|x\|} \cdot n \right) n = \left[ \frac{f}{\|x\|} \otimes n - n \otimes \frac{f}{\|x\|} \right] n, \tag{6}
\]

where \( u \otimes y \) denotes the dyadic operation of \( u \) and \( y \), i.e., \( (u \otimes y)z = y \cdot zu \).

By the definition of \( x = \|x\|n \), and from Eqs. (4)-(6), we can derive

\[
\dot{x} = \frac{d}{dt} \|x\| n + \|x\| \dot{n} = \frac{f \cdot x}{\|x\|^2} x + \left[ \frac{f}{\|x\|} \otimes \frac{x}{\|x\|} - \frac{x}{\|x\|} \otimes \frac{f}{\|x\|} \right] x, \tag{7}
\]

and at the same time Eq. (5) can be written as

\[
\frac{d}{dt} \|x\| = f \cdot \frac{x}{\|x\|}. \tag{8}
\]

Then, Eqs. (7) and (8) can be put together as

\[
\frac{d}{dt} \begin{bmatrix} x \\ \|x\| \end{bmatrix} = \begin{bmatrix} \frac{f \otimes x}{\|x\|} - \frac{x}{\|x\|} \otimes \frac{f}{\|x\|} & \frac{(f \cdot x)x}{\|x\|^3} \\ \frac{(f \cdot x)x^T}{\|x\|^3} & 0 \end{bmatrix} \begin{bmatrix} x \\ \|x\| \end{bmatrix}. \tag{9}
\]

Furthermore, in terms of

\[
X := \begin{bmatrix} x \\ \|x\| \end{bmatrix}, \quad B := \begin{bmatrix} \frac{f \otimes x}{\|x\|} - \frac{x}{\|x\|} \otimes \frac{f}{\|x\|} & \frac{(f \cdot x)x}{\|x\|^3} \\ \frac{(f \cdot x)x^T}{\|x\|^3} & 0 \end{bmatrix}, \tag{10}
\]

and in terms of the symmetric and skew-symmetric matrices:

\[
S = \frac{f \cdot x}{\|x\|^2} I_n, \quad W = \frac{f}{\|x\|} \otimes \frac{x}{\|x\|} - \frac{x}{\|x\|} \otimes \frac{f}{\|x\|}, \tag{11}
\]

we can write Eq. (9) as a Lie-type ODEs:

\[
\dot{X} = BX, \tag{12}
\]

where

\[
B = \begin{bmatrix} W & S \\ (Sn)^T & 0 \end{bmatrix}. \tag{13}
\]
We can find three structures about Eq. (9):

\[
\begin{align*}
\text{Cone:} & \quad X^T g X = 0, \\
\text{Lie-algebra:} & \quad B \in so(n,1), \quad B^T g + gB = 0, \\
\text{Lie-group:} & \quad G \in SO_o(n,1), \quad G^T g G = g,
\end{align*}
\]

where \( g := \text{diag}(I_n, -1) \) is the metric tensor of the Minkowski space \( \mathbb{M}^{n+1} \) with a signature \((n, 1)\), and \( G \) is the Lie-group generated from \( B \) [Liu (2001)]. Recently, Liu (2013a) has developed a Lie-group algorithm based on the Lie-symmetry \( GL(n, \mathbb{R}) \) by dropping out the first and the third terms in the right-hand side of Eq. (7). Moreover, Liu (2013b) has developed a Lie-group algorithm based on the Lie-symmetry \( DSO(n) \) by Eq. (7). In Fig. 1 we show the relations of those Lie-group formulations to Eq. (3). The present formulation by using the full Lie-symmetry of \( SO_o(n, 1) \) is more delicate, as to be shown below.

### 3 A signum function

In order to develop a numerical scheme from Eq. (9), we suppose that the coefficient matrix is constant with the pair

\[
(a, b) = \left( \frac{\ddot{f}}{\|\ddot{x}\|}, \frac{\ddot{x}}{\|\ddot{x}\|} \right)
\]

being constant, which can be obtained by taking the values of \( \ddot{f} \) and \( \ddot{x} \) at a suitable mid-point of \( \ddot{t} \in [0, t] \), where \( t \leq h \) and \( h \) is a small time stepsize.

From Eqs. (9) and (17) we thus need to solve a constant linear system:

\[
\frac{d}{dt} \begin{bmatrix} x \\ \|x\| \end{bmatrix} = \begin{bmatrix} a \otimes b - b \otimes a & a \cdot bb^T \\ a \cdot bb^T & 0 \end{bmatrix} \begin{bmatrix} x \\ \|x\| \end{bmatrix}.
\]

Let

\[
\begin{align*}
z & = a \cdot x, \quad w = b \cdot x, \quad y = \|x\|,
\end{align*}
\]

\[
\begin{align*}
c_0 & = a \cdot b,
\end{align*}
\]

and Eq. (18) becomes

\[
\begin{align*}
\dot{x} & = wa - zb + c_0yb, \\
\dot{y} & = c_0w.
\end{align*}
\]
At the same time, from the above equations we can derive the following ODEs for \( z, w \) and \( y \):

\[
\frac{d}{dt} \begin{pmatrix} z \\ w \\ y \end{pmatrix} = \begin{pmatrix} -c_0 & a_0^2 & c_0^2 \\ -1 & c_0 & 0 \\ 0 & c_0 & 0 \end{pmatrix} \begin{pmatrix} z \\ w \\ y \end{pmatrix},
\]

(23)

where \( a_0 = \|a\| \). Fortunately, the original \((n+1)\)-dimensional problem in Eq. (18) can be reduced to a three-dimensional problem in the above.

For the special case with \( c_0 = 0 \) we can derive

\[
x(t) = x_0 + \left[ \frac{[\cos(\Omega t) - 1]z_0}{\Omega^2} + \frac{\sin(\Omega t)w_0}{\Omega} \right] a + \left[ \cos(\Omega t) - 1 \right] a_0^2 + \frac{\sin(\Omega t)z_0}{\Omega} b,
\]

(24)

where \( \Omega = \|a\| \), \( z_0 = a \cdot x_0 \) and \( w_0 = b \cdot x_0 \). For this case \( \|x\| \) is a constant.

Here we give a detailed derivation of the solutions for \((z, w, y)\) with \( c_0 \neq 0 \). Depending on the signum function of \( \text{sign}(a_0^2 - 2c_0^2) \), there exist two different types solutions of \((z, w, y)\) of Eq. (23). Instead of \( \text{sign}(a_0^2 - 2c_0^2) \), we will use \( \text{Sign} \) for a shorthand.

From the first-order ODEs in Eq. (23) we have

\[
\dot{z} = a_0^2 w + c_0^2 y - c_0 z,
\]

\[
\dot{w} = c_0 w + c_0 y - z,
\]

\[
\dot{y} = c_0 w.
\]

(26)

It follows that

\[
\frac{d^3 y}{dt^3} + (a_0^2 - 2c_0^2) \dot{y} = 0.
\]

(27)

Depending on the value of \( a_0^2 - 2c_0^2 \), \( y \) has two different types of solutions.

(A) For the first case with \( a_0^2 - 2c_0^2 < 0 \), i.e., \( \text{Sign} = -1 \), we have

\[
y(t) = k_0 + k_1 h_1(t) + k_2 h_2(t),
\]

(28)

where \( k_0, k_1 \) and \( k_2 \) are constants and

\[
h_1(t) = \sinh(\Omega t), \quad h_2(t) = \cosh(\Omega t),
\]

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in which \( \Omega = \sqrt{2c_0^2 - a_0^2} \).

Through some elementary operations we can derive the following solutions for \( z, w \) and \( y \):

\[
P = \begin{pmatrix}
c_2 + d_1h_1 + c_3h_2 & d_2h_1 + d_3h_2 - c_0c_2 & c_2\Omega h_1 + c_1c_2(h_2 - 1) \\
-\frac{c_2\Omega h_1}{c_0} & \frac{c_3\Omega h_1}{c_0} + h_2 & \frac{c_3\Omega h_1}{c_0} + c_2(h_2 - 1) \\
c_2(1-h_3) & \frac{c_0h_1}{c_0} + c_2(h_2 - 1) & c_3 + c_2h_2(t)
\end{pmatrix},
\]

where \( z_0, w_0 \) and \( y_0 \) are initial values of \( z, w \) and \( y \) at an initial time \( t = 0 \), and

\[
\Omega = \sqrt{2c_0^2 - a_0^2}; \quad c_1 = \frac{c_0^2 - \Omega^2}{c_0}, \quad c_2 = \frac{c_0}{c_0 - c_1}, \quad c_3 = -\frac{c_1c_2}{c_0};
\]

\[
d_1 = -\frac{c_2\Omega}{c_0}, \quad d_2 = \frac{c_0c_1}{\Omega} + c_2\Omega, \quad d_3 = c_0 + c_1c_2.
\]

(B) For the second case with \( a_0^2 - 2c_0^2 > 0 \), i.e., \( \text{Sign} = +1 \), we have

\[
y(t) = k_0 + k_1h_3(t) + k_2h_4(t),
\]

where \( k_0, k_1 \) and \( k_2 \) are constants and

\[
h_3(t) = \sin(\Omega t), \quad h_4(t) = \cos(\Omega t),
\]

in which \( \Omega = \sqrt{a_0^2 - 2c_0^2} \).

Through some elementary operations we can derive the following solutions for \( z, w \) and \( y \):

\[
Q = \begin{pmatrix}
c_2 + d_1h_3 + c_3h_4 & d_2h_3 + d_3h_4 - c_0c_2 & c_1c_2(h_4 - 1) - c_2\Omega h_3 \\
\frac{c_2\Omega h_1}{c_0} & \frac{h_4 - c_2\Omega h_3}{c_0} & -\frac{c_2\Omega h_1}{c_0} \\
c_2(1-h_4) & \frac{c_0h_3}{c_0} + c_2(h_4 - 1) & c_3 + c_2h_4
\end{pmatrix},
\]

where

\[
\Omega = \sqrt{a_0^2 - 2c_0^2}; \quad c_1 = \frac{c_0^2 + \Omega^2}{c_0}, \quad c_2 = \frac{c_0}{c_0 - c_1}, \quad c_3 = -\frac{c_1c_2}{c_0};
\]

\[
d_1 = \frac{c_2\Omega}{c_0}, \quad d_2 = \frac{c_0c_1}{\Omega} - c_2\Omega, \quad d_3 = c_0 + c_1c_2.
\]
When \((z, w, y)\) are obtained we can insert them into Eq. (21) and integrate the resultant equation to obtain the solution of \(x(t)\). The solutions are written out directly.

\[
\text{sign}(c_0) = +1\quad \text{sign}(c_0) = -1
\]

![Figure 2: A schematic plot of the area of Sign=+1, Sign=-1, sign(c_0) = +1, and sign(c_0) = -1 in the plane.](image)

(A) For the case with \(a_0^2 - 2c_0^2 < 0\) (Sign = -1), inserting Eq. (30) for \(z, w\) and \(y\) into Eq. (21) and integrating the resultant equation we can obtain

\[
x(t) = x_0 + [G_1(t)z_0 + G_2(t)w_0 + G_5(t)||x_0||a + [G_3(t)z_0 + G_4(t)w_0 + G_6(t)||x_0||b],
\]

where

\[
\begin{align*}
G_1(t) &= \frac{c_2[1 - h_2(t)]}{c_0^2}, \\
G_2(t) &= \frac{c_2[h_2(t) - 1]}{c_0} + \frac{h_1(t)}{\Omega}, \\
G_3(t) &= \frac{d_1[1 - h_2(t)]}{\Omega} - \frac{(c_2 + c_3)h_1(t)}{\Omega}, \\
G_4(t) &= \frac{(c_0^2 - d_2\Omega)[h_2(t) - 1]}{\Omega^2} + \frac{(c_0c_2 - d_3)h_1(t)}{\Omega}, \\
G_5(t) &= \frac{c_2[h_2(t) - 1]}{c_0}, \\
G_6(t) &= \frac{c_0h_1(t)}{\Omega} - c_2[h_2(t) - 1],
\end{align*}
\]
in which \( \Omega = \sqrt{2c_0^2 - a_0^2} \), \( h_1(t) = \sinh(\Omega t) \), and \( h_2(t) = \cosh(\Omega t) \).

(B) Next for the case with \( a_0^2 - 2c_0^2 > 0 \) (Sign = +1) we can obtain

\[
x(t) = x_0 + [g_1(t)z_0 + g_2(t)w_0 + g_5(t)\|x_0\|]a \\
+ [g_3(t)z_0 + g_4(t)w_0 + g_6(t)\|x_0\|]b,
\]

where

\[
g_1(t) = \frac{c_2[1-h_4(t)]}{c_0^2}, \\
g_2(t) = \frac{c_2[h_4(t)-1]}{c_0} + \frac{h_3(t)}{\Omega}, \\
g_3(t) = \frac{d_1[h_4(t)-1]}{\Omega} - \frac{(c_2 + c_3)h_3(t)}{\Omega}, \\
g_4(t) = \frac{(c_0^2 - d_2\Omega)[1-h_4(t)]}{\Omega^2} + \frac{(c_0c_2 - d_3)h_3(t)}{\Omega}, \\
g_5(t) = \frac{c_2[h_4(t)-1]}{c_0}, \\
g_6(t) = \frac{c_0h_3(t)}{\Omega} - c_2[h_4(t)-1],
\]

in which \( \Omega = \sqrt{a_0^2 - 2c_0^2} \), \( h_3(t) = \sin(\Omega t) \), and \( h_4(t) = \cos(\Omega t) \).

4 Numerical algorithm: GPS2

From the above solutions of \( x(t) \) we can derive the new algorithm. In order to distinct the present method from the group-preserving scheme (GPS) developed by Liu (2001), we may call the new algorithm to be the second GPS, denoted by GPS2, which is summarized as follows.

(i) Give an initial value of \( x_0 \) at an initial time \( t = t_0 \) and a time stepsize \( h \).

(ii) For \( k = 0, 1, \ldots \), we repeat the following computations to a specified terminal
Figure 3: For the Duffing oscillator under a harmonic loading, showing (a) the signum function and (b) the length.

time $t = t_f$:

\[ \mathbf{a}_k = \frac{f_k}{\|x_k\|}, \]

\[ \mathbf{b}_k = \frac{x_k}{\|x_k\|}, \]

\[ a^0_k = \|\mathbf{a}_k\|, \]

\[ c^0_k = \mathbf{a}_k \cdot \mathbf{b}_k, \]

\[ z^k_0 = \mathbf{a}_k \cdot x_k, \]

\[ w^k_0 = \mathbf{b}_k \cdot x_k, \]

\[ \text{Sign} = \frac{(a^0_k)^2 - 2(c^0_k)^2}{|(a^0_k)^2 - 2(c^0_k)^2|}, \]
\[
\begin{align*}
\Omega_k &= \sqrt{(a_0^k)^2 - 2(c_0^k)^2}, \\
c_1^k &= \frac{(c_0^k)^2 + \Omega_k^2}{c_0^k}, \\
c_2^k &= \frac{c_0^k}{c_0^k - c_1^k}, \\
c_3^k &= -\frac{c_1^k}{c_0^k}, \\
d_1^k &= \frac{c_0^k}{\Omega_k}, \\
d_2^k &= \frac{c_0^k}{\Omega_k} - c_1^k \Omega_k, \\
d_3^k &= c_0^k + c_1^k c_2^k, \\
a_k &= \sin(\Omega_k h), \\
b_k &= \cos(\Omega_k h), \\
g_3^k &= \frac{d_1^k (b_k - 1) - (c_0^k + c_1^k) a_k}{\Omega_k}, \\
g_4^k &= \frac{[(c_0^k)^2 - d_1^k \Omega_k](1-b_k)}{\Omega_k^2} + \frac{(c_0^k c_1^k - d_1^k) a_k}{\Omega_k}.
\end{align*}
\]

if \( \text{Sign} = +1 \),

\[
\begin{align*}
\Omega_k &= \sqrt{2(c_0^k)^2 - (a_0^k)^2}, \\
c_1^k &= \frac{(c_0^k)^2 - \Omega_k^2}{c_0^k}, \\
c_2^k &= \frac{c_0^k}{c_0^k - c_1^k}, \\
c_3^k &= -\frac{c_1^k}{c_0^k}, \\
d_1^k &= -\frac{c_0^k}{\Omega_k}, \\
d_2^k &= \frac{c_0^k}{\Omega_k} + c_1^k \Omega_k, \\
d_3^k &= c_0^k + c_1^k c_2^k, \\
a_k &= \sinh(\Omega_k h), \\
b_k &= \cosh(\Omega_k h), \\
g_3^k &= \frac{d_1^k (1-b_k) - (c_0^k + c_1^k) a_k}{\Omega_k}, \\
g_4^k &= \frac{[(c_0^k)^2 - d_1^k \Omega_k](1-b_k)}{\Omega_k^2} + \frac{(c_0^k c_1^k - d_1^k) a_k}{\Omega_k}.
\end{align*}
\]

if \( \text{Sign} = -1 \),

\[
\begin{align*}
g_1^k &= \frac{c_2^k (1-b_k)}{(c_0^k)^2}, \\
g_2^k &= \frac{c_2^k (b_k - 1)}{c_0^k} + \frac{a_k}{\Omega_k}, \\
g_3^k &= \frac{c_2^k (b_k - 1)}{c_0^k}, \\
g_5^k &= \frac{c_0^k a_k}{\Omega_k} - c_2^k (b_k - 1), \\
x_{k+1} &= x_k + (g_1^k x_k + g_2^k w_0^k + g_5^k \|x_k\|)a_k + (g_3^k x_k + g_4^k w_0^k + g_6^k \|x_k\|)b_k.
\end{align*}
\]
In Eq. (40) there are \( \sin(\Omega_k h) \) and \( \cos(\Omega_k h) \), while that in Eq. (41) there are \( \sinh(\Omega_k h) \) and \( \cosh(\Omega_k h) \). The most dynamical systems fall into the first class with \( \text{Sign} = +1 \). However, for a chaotic system the situation is quite different, where both equations (40) and (41) are needed in the computation.

5 The barcode and the set \( \mathcal{A}_1^- \)

It is significant that in Eq. (25) we have derived a signum function, which is abbreviated as \( \text{Sign} \) for saving notation, to demand the algorithm into two classes in Eqs. (40) and (41). Without having the factor 2 before \( \mathbf{f} \cdot \mathbf{x} \) in Eq. (25) one has

\[
\|\mathbf{f}\|^2\|\mathbf{x}\|^2 - (\mathbf{f} \cdot \mathbf{x})^2 = \|\mathbf{f}\|^2\|\mathbf{x}\|^2 (1 - \cos^2 \theta) \geq 0, 
\]

where \( \theta \) is the intersection angle between \( \mathbf{x} \) and \( \mathbf{f} \); hence, it makes no sense to say its signum function, because it is always non-negative.

On the contrast, by Eq. (25) we have

\[
\text{Sign} = \text{sign}(\|\mathbf{f}\|^2\|\mathbf{x}\|^2 (1 - 2\cos^2 \theta)) = -\text{sign}(\cos 2\theta),
\]

which might be +1 or −1, depending on the intersection angle \( \theta \) between \( \mathbf{x} \) and \( \mathbf{f} \). When \( \theta \) is in the range of \( -\pi/2 < \theta < \pi/2 \) or \( 3\pi/4 < \theta < 5\pi/4 \), the value of \( \text{Sign} \) is \( \text{Sign} = -1 \). As a remark given below Eq. (23), here we can classify the behavior of the nonlinear dynamical system in a three-dimensional subspace \( (\mathbf{f}, \mathbf{x}, \|\mathbf{x}\|) \) with a trio \( (\text{Sign}, \text{sign}(c_0), \|\mathbf{x}\|) \). Thus we can observe the time-varying values of \( \text{Sign} \) and plot them as a barcode with alternative values of \( \text{Sign} \) being +1 and -1. It is known that a barcode is an optical machine-readable representation of data relating to the object to which it is attached. A main feature of the barcode is the intervened black lines and white lines with varying spacings and widths. Barcode is ubiquitously used in the identification and classification of products. Here we will use the barcode to identify the property of nonlinear oscillators in Eqs. (1) and (2).

**Definition 1:** The barcode is a time-varying function of \( \text{Sign} \) defined in a time interval.

Let us give a schematic plot in Fig. 2, and as shown there we have two dis-connected sets of \( \|\mathbf{f}\|^2\|\mathbf{x}\|^2 - 2(\mathbf{f} \cdot \mathbf{x})^2 < 0 \):

\[
\mathbf{f} \cdot \mathbf{x} > \frac{1}{\sqrt{2}}\|\mathbf{f}\|\|\mathbf{x}\|,
\]

\[
\mathbf{f} \cdot \mathbf{x} < -\frac{1}{\sqrt{2}}\|\mathbf{f}\|\|\mathbf{x}\|,
\]

(44) 

(45)
by noting that \((x, f) = (0, 0)\) is deleted. Then we can identify five time-varying sets:

\[
\mathcal{A}_1^- := \{(x, f) \mid \text{Sign} = -1, \text{ and } f \cdot x > \frac{1}{\sqrt{2}}\|f\|\|x\|\}
\]

\[
= \{(x, f) \mid \|f\|^2\|x\|^2 - 2(f \cdot x)^2 < 0, \text{ and } f \cdot x > \frac{1}{\sqrt{2}}\|f\|\|x\|\},
\]

(46)

\[
\mathcal{A}_2^- := \{(x, f) \mid \text{Sign} = -1, \text{ and } f \cdot x < -\frac{1}{\sqrt{2}}\|f\|\|x\|\}
\]

\[
= \{(x, f) \mid \|f\|^2\|x\|^2 - 2(f \cdot x)^2 < 0, \text{ and } f \cdot x < -\frac{1}{\sqrt{2}}\|f\|\|x\|\},
\]

(47)

\[
\mathcal{A}^+ := \{(x, f) \mid \text{Sign} = +1\} = \{(x, f) \mid \|f\|^2\|x\|^2 - 2(f \cdot x)^2 > 0\},
\]

(48)

\[
\mathcal{B}^+ := \{(x, f) \mid \text{sign}(c_0) = +1\} = \{(x, f) \mid f \cdot x > 0\},
\]

(49)

\[
\mathcal{B}^- := \{(x, f) \mid \text{sign}(c_0) = -1\} = \{(x, f) \mid f \cdot x < 0\}.
\]

(50)

While \(\mathcal{B}^+\) and \(\mathcal{B}^-\) are connected, \(\mathcal{A}_1^-\) and \(\mathcal{A}_2^-\) are dis-connected. We may call \(\mathcal{A}_1^-\) the first set of dis-connectivity and \(\mathcal{A}_2^-\) the second set of dis-connectivity, respectively. Clearly, the first set of dis-connectivity is a subset of \(\mathcal{B}^+\) and the second set of dis-connectivity is a subset of \(\mathcal{B}^-\), i.e.,

\[
\mathcal{A}_1^- \subset \mathcal{B}^+, \quad \mathcal{A}_2^- \subset \mathcal{B}^-.
\]

(51)

We can prove the following theorems, which can help us understand the complex structure of a barcode.

**Theorem 1:** If Eq. (3) satisfies

\[
f(x(t_0), t_0) \cdot x(t_0) > 0, \text{ and } \text{Sign} = -1 \, \forall t \geq t_0,
\]

then \(x(t) \to \infty, \, t \to \infty.\)  \hspace{1cm} (52)

**Proof:** Under the first assumption the case in Eq. (45) is impossible because it contradicts to \(f(x(t_0), t_0) \cdot x(t_0) > 0.\) Then under the condition of \(\text{Sign} = \text{sign}(\|f\|^2\|x\|^2 - 2(f \cdot x)^2) = -1,\) it is always

\[
f \cdot x > \frac{1}{\sqrt{2}}\|f\|\|x\| \, \forall t \geq t_0,
\]

(53)

which is in the set \(\mathcal{A}_1^-\) by Eq. (46). Because the two sets \(\mathcal{A}_1^-\) and \(\mathcal{A}_2^-\) are dis-connected, and from the first set \(\mathcal{A}_1^-\) to the second set \(\mathcal{A}_2^-\) it must be \(\text{Sign} = \text{sign}(\|f\|^2\|x\|^2 - 2(f \cdot x)^2) = +1\) on some time interval, which contradicts to the second assumption in Eq. (52). Then using Eqs. (8) and (53) we have

\[
\frac{d}{dt}\|x\| > \frac{1}{\sqrt{2}}\|f\| > 0,
\]

(54)
which means that the length grows with time. Thus, Eq. (52) is proven. □

**Theorem 2:** If Eq. (3) satisfies

\[
    f(x(t_0), t_0) \cdot x(t_0) < 0, \quad \text{and} \quad \text{Sign} = -1 \quad \forall t \geq t_0,
\]

then \( x(t) \to 0, \quad t \to \infty \). \hfill (55)

**Proof:** Under the first assumption the case in Eq. (44) is impossible because it contradicts to \( f(x(t_0), t_0) \cdot x(t_0) < 0 \). Then under the condition of \( \text{Sign} = \text{sign}(\|f\|\|x\|^2 - 2(f \cdot x)^2) = -1 \), it is always

\[
    f \cdot x < -\frac{1}{\sqrt{2}} \|f\|\|x\| \quad \forall t \geq t_0,
\]

which is in the set \( A^{-2} \) by Eq. (47). By a similar argument and then using Eqs. (8) and (56) we have

\[
    \frac{d}{dt}\|x\| < -\frac{1}{\sqrt{2}} \|f\| < 0,
\]

which means that the length decreases with time. Thus, Eq. (55) is proven. □

We note that the length as governed by Eq. (8) has the following property:

\[
    \text{sign}(c_0) \leq 0 \Leftrightarrow \frac{d}{dt}\|x\| \leq 0,
\]

where

\[
    \text{sign}(c_0) = \text{sign}(f \cdot x) = \text{sign}(\cos \theta).
\]

The above two theorems also hold if we replace \( A^{-1} \) and \( A^{-2} \) by the sets \( B^+ \) and \( B^- \), but they are un-interesting, because for an oscillatory system the value of \( \text{sign}(c_0) \) always changes its sign between +1 and -1.

For a chaotic system the value of Sign is not always +1 or -1. In order to demonstrate the use of the barcode, let us investigate the Duffing oscillator (1) under the following parameters: \( \gamma = 0.3, \quad \alpha = -1, \quad \beta = 1, \quad F_0 = 0.32, \quad \omega = 1.2 \). In all computations the initial conditions are set to be \( (x(0), \dot{x}(0)) = (0, 0) \), unless specifying otherwise. By using the algorithm GPS2 in Section 4 to solve the Duffing equation the time stepsize is \( h = 0.01 \). We analyze the signum function and the length as shown in Fig. 3 within a time interval of \( t \in [200, 220] \), where the transient part in \( t \in [0, 200] \) is not plotted for clear.
As shown in Fig. 3(a) by solid line the values of Sign are varying from $-1$ to $+1$ and then to $-1$, and then Sign will return to $+1$ again; otherwise, as shown in Theorem 1 the system will respond unstably, causing the displacement tend to infinity, which by definition is not a chaotic system. In order to compare the values of $\text{sign}(c_0)$ with the values of Sign we plot them in Fig. 3(a) with dashed and solid line, respectively, where we make a slight shift of the dashed line downward for clear. On the other hand, for a distinction to $A_{-2}$, the first set of dis-connectivity $A_{-1}$ is filled by solid black points.

**Theorem 3:** If a state is in the set $A_{-1} \subset B^+$, before it leaves the set $B^+$ to $B^-$, the set $A_{-1}$ changes to $A^+$. 

**Proof:** It follows from the first equation in Eq. (51) and the dis-connectivity of $A_{-1}$ and $A_{-2}$. □

**Theorem 4:** If a state is in the set $A_{-2} \subset B^-$, before it leaves the set $B^-$ to $B^+$, the set $A_{-2}$ changes to $A^+$. 

**Proof:** It follows from the second equation in Eq. (51) and the dis-connectivity of $A_{-1}$ and $A_{-2}$. □

Before $\text{sign}(c_0)$ changes from $+1$ to $-1$, for example from point $f_1$ to point $f_2$ in Fig. 2, the Sign must jump from $-1$ to $+1$. The two jumping behaviors described in Theorems 3 and 4 render a quite complex structure of the barcode as shown in Fig. 3(a). More barcodes are to be shown below, which can be seen very complex. From Fig. 3(a) we can observe the following interesting phenomena: (i) The first set of dis-connectivity $A_{-1}$ is a subset of $\text{sign}(c_0) = +1$, and the second set of dis-connectivity $A_{-2}$ is a subset of $\text{sign}(c_0) = -1$. (ii) In a time interval of $\text{Sign} = -1$, the state is either in the first set of dis-connectivity $A_{-1}$ or in the second set of dis-connectivity $A_{-2}$, which is due to the dis-connectivity of $A_{-1}$ and $A_{-2}$. (iii) From one in first set of dis-connectivity $A_{-1}$ to another first set of dis-connectivity $A_{-1}$ there must accompany a jump from $\text{Sign} = -1$ to $\text{Sign} = +1$. This also holds for the second set of dis-connectivity $A_{-2}$. (iv) From one in first set of dis-connectivity $A_{-1}$ to one in second set of dis-connectivity $A_{-2}$, or vice-versa, there must accompany a jump from $\text{Sign} = -1$ to $\text{Sign} = +1$. For some cases this jumping event only happens at one time point [an example will be given in Fig. 11(b)]. (v) Before $\text{sign}(c_0)$ jumps from $+1$ to $-1$ as remarked in Fig. 3(a) by the symbols $+$ and $-$ (i.e., the length is decreased), and if Sign is in the state of $-1$, then Sign will jump
from $-1$ to $+1$ as remarked in Fig. 3(a) by the symbols $-$ and $+$. The proof of (v) is obvious by viewing Fig. 2 and that the first set of $\text{Sign} = -1$ cannot directly jump to the second set of $\text{Sign} = -1$. When $f_1$ goes to $f_2$, the $\text{Sign}$ changes from $-1$ to $+1$.

Moreover, according to the numerical algorithm of GPS2 we can prove the following result.

**Theorem 5:** If Eq. (3) satisfies

$$\text{Sign} = +1 \quad \forall t \geq t_0,$$

then $x(t)$ is stable,

$$x(t)$$

where $t_0$ is a small finite time.

**Proof:** Under the above condition of $\text{Sign} = +1$, the algorithm in Eq. (40) is used. Because $|\sin(\Omega_k h)| \leq 1$ and $|\cos(\Omega_k h)| \leq 1$, the algorithm does not bring the state of $x$ to be unbounded. Indeed, the Lie-group transformation from $x_k$ to $x_{k+1}$ is a compact subgroup of $SO_o(n, 1)$, whose action is bounded. □

Theorems 1 and 2 reflect two extremal cases with the state always being $\text{Sign} = -1$, which is either unstable or tending to zero. Thus for a regular system it should go to the state $\text{Sign} = +1$ after a certain time $t_0$ as shown in Theorem 5. However, a chaotic system is frequently switching between these two states of $\text{Sign} = +1$ and $\text{Sign} = -1$. This is the reason that the barcode of a chaotic system is quite complex. In the next section we will demonstrate the importance of the first set of dis-connectivity $\mathcal{A}_f$ for Duffing equation (1) and the ship rolling equation (2).

### 6 Results and discussions

In this section we apply the GPS2 to solve Eqs. (2) and (1), and investigate their behaviors by using the barcode and its first set of dis-connectivity $\mathcal{A}_f$.

**(A) Ship rolling equation:**

(i) For Eq. (2) with $\gamma = 0.1$ (we fix it, unless specifying otherwise), under a fixed $F_0 = 0.2$, and with two different $\omega = 0.76$ and $\omega = 0.6$, we plot the phase portraits of $(x, y) = (x, \dot{x})$ in Fig. 4(a), where we can observe two type behaviors: The escape one with $\omega = 0.76$ whose Sign function is plotted in Fig. 4(b) with an unstable behavior as specified by Theorem 1. Here we can only calculate the solution up to $t = 12$ sec although we use a very small time stepsize $h = 0.0005$. The periodic one (in steady state) with $\omega = 0.6$, whose Sign function is plotted in Fig. 4(c) with
stable behavior as specified by Theorem 5. The Sign is returned to Sign = +1 after 25 sec and until 200 sec it keeps Sign = +1 unchanged.

(ii) In order to further test the escape and chaotic behavior of Eq. (2), we plot the percentage of the state in the first set of dis-connectivity $A^-$ up to 500 sec for two fixed values, $F_0 = 0.2$ in Fig. 5(a) and $F_0 = 0.15$ in Fig. 5(b), with respect to the excitation frequency $\omega$ in the range of $0.1 \leq \omega \leq 1$. The curves are fitted with black points and connected by dashed lines and in the escape state there exist no black points. The GPS2 has a mechanism to identify the "instability" of solution by checking the value of $\Omega_k h$ in Eq. (41). Because $\Omega_k h$ appears in $\cosh(\Omega_k h)$ and $\sinh(\Omega_k h)$, when $\Omega_k h \geq 20$ the algorithm will blow up and we terminate the computation, although the time does not reach to $t = 500$ sec, of which the resultant state is classified as "escape" in the parameter space of $\omega$. Thompson, Bishop and Leung (1987) have employed $x > 20$ to be an escape criterion. Here we employ the criterion of $\Omega_k h \geq 20$, basing on the numerical algorithm. For the case of $F_0 = 0.2$, after $\omega > 0.6$ there are all escape states, and another one indicated in Fig. 5(a) by a piece of dashed line. After the escape state there exists a thin layer within which the motion is chaotic. In the percentage curve of $A^-$ there appears many peaks, such as those marked by the numbers 0.2935, 0.334 and 0.5635 in Fig. 5(a), and under these parameter values of $\omega$ we find respectively, 1/3 subharmonic, 1/2 subharmonic and periodic motion as shown in the inset. In the chaotic state, the percentage of $A^-$ is very low, smaller than 1%, for the ship rolling equation (2). We remind that in the hyperplane $(x,f)$ as shown in Fig. 2, $A^-$ might have the maximal percentage 25%.

(iii) In order to show the chaotic behavior of Eq. (2), we compare two steady state behaviors in Fig. 6(a) with the same $\omega = 0.17$ but with two slightly different values of $F_0 = 0.2414$ and $F_0 = 0.2415$, which correspond respectively to 1/4 subharmonic and chaotic motion. The corresponding barcodes as compared in Fig. 6(b) for $F_0 = 0.2415$, and Fig. 6(c) for $F_0 = 0.2414$ are slightly different. Although, the difference of $F_0$ is only 0.0001, but the resultant behaviors are very different. This indicates that Eq. (2) is a chaotic system under certain parameter values, which is sensitive to the parameter value.

(iv) In Figs. 7(a) and 7(b) we fix $\gamma = 0.1$ and $\gamma = 0.05$ by plotting the distribution of escape states in the parameter space of $(\omega, F_0) \in [0, 1] \times [0, 0.4]$, where the escape states are marked by black points. For the purpose of comparison the following Melnikov curve [Thompson (1989)] is plotted in Fig. 7:

$$F_M = \frac{\gamma \sinh(\pi \omega)}{5\pi \omega^2}. \quad (61)$$

$F_0 > F_M$ is a necessary condition for the appearance of chaos, but it is not a sufficient condition. In Fig. 8 we plot the escape time which is the terminal time when
\( \Omega_k h \) used in in Eq. (41) is arrived to 20. After that the algorithm blows up. There have several cusps and adjunct them the state of chaos appears. For example the above case with \( F_0 = 0.2415 \) is near to a cusp of escape. In the largest cusp region there is a fractal structure as shown in the inset of Fig. 7(a), which is plotted in the range of \((\omega, F_0) \in [0.8, 1] \times [0.06, 0.16] \).

Figure 4: For a quadratic oscillator under a harmonic loading, (a) escape and chaotic orbit, and the signum functions for (b) escape motion and (c) periodic motion.
Figure 5: For quadratic oscillators under harmonic loadings, showing escape and chaotic regions for two amplitudes of force. The inset shows subharmonic motions.

(v) Within the large cusp, Thompson, Rainey and Soliman (1990) have analyzed the bifurcation behavior with a fixed value $\omega = 0.85$ and varying $F_0 \in [0.107, 0.109]$. However, in that range we cannot detect any special structure by using the new theory. In Fig. 9 with $\omega = 0.85$ being fixed, we plot the bifurcation diagram by using the percentage curve of $A^\perp_1$ with respect to $F_0$ in the range of $F_0 \in [0.06, 0.108]$. 
Figure 6: For quadratic oscillators, (a) showing subharmonic and chaotic motions with slightly different amplitudes of force, and (b) and (c) the corresponding barcodes.
Figure 7: For quadratic oscillators under harmonic loadings, showing escape regions for two damping constants.
Figure 8: For quadratic oscillators under harmonic loadings, showing escape times for two damping constants: (a) $\gamma = 0.1$, and (b) $\gamma = 0.05$.
It is interesting that after a sequence of period-doubling bifurcation showing as a devil staircase in the figure there are states of escape and chaos.

Figure 9: For quadratic oscillators under fixed $\omega = 0.85$ and varying amplitudes of loading, showing a bifurcation diagram with the staircase structure.

(B) Duffing equation:

(i) For the Duffing Eq. (1) we fix the parameters as those given in the computation of Fig. 3. Under the same parameters of $\gamma = 0.3$, $\alpha = -1$, $\beta = 1$, $\omega = 1.2$, we plot the curves of the percentages of sign($c_0$) = +1 (i.e., $B^+$) and $A^-$ with respect to $F_0$ from $F_0 = 0.22$ to $F_0 = 0.4$ in Fig. 10. The time is up to 250 sec and with $h = 0.005$ sec used in the GPS2. As expected the curve in Fig. 10(a) gives no much useful information about the motion types, whose percentages are near 50%. However, in the percentage curve of $A^-$ in Fig. 10(b) there appears many peaks, such as those marked by the numbers 0.267, 0.288 and 0.301, 0.356 and 0.359, and under these parameter values of $\omega$ we find respectively, a smaller 1/3 subharmonic (Fig. 11),
Figure 10: For the Duffing oscillator under different amplitudes of harmonic loading, showing the percentage of (a) the signum function $c_0$, and (b) the first set of dis-connectivity.
Figure 11: For the Duffing oscillator under $F_0 = 0.267$, showing (a) the steady orbit of 1/3 subharmonic motion, and (b) the barcode.
Figure 12: For the Duffing oscillator under $F_0 = 0.288$, showing (a) the steady orbit of $1/4$ subharmonic motion, and (b) the barcode.
Figure 13: For the Duffing oscillator under $F_0 = 0.301$, showing (a) the steady orbit of chaotic motion, and (b) the barcode.
Figure 14: For the Duffing oscillator under $F_0 = 0.356$, showing (a) 1/3 subharmonic motion inside a window of chaotic range, and (b) the barcode.
Figure 15: For the Duffing oscillator under $F_0 = 0.359$, showing (a) 1/5 subharmonic motion inside a window of chaotic range, and (b) the barcode.
Figure 16: For the Duffing oscillator under $F_0 = 0.2895$, comparing the results of (a) RK4, and (b) GPS2. RK4 gives an incorrect 1/4 subharmonic motion.
1/4 subharmonic (Fig. 12), chaotic (Fig. 13), a larger 1/3 subharmonic (Fig. 14) and a larger 1/5 subharmonic (Fig. 15). The last two subharmonic motions are within the subharmonic window inside the chaotic range. The range of 1/3 subharmonic is \( \omega \in [0.356, 0.359) \), while the range of 1/5 subharmonic is \( \omega \in [0.359, 0.38) \). After \( \omega = 0.38 \) the Duffing system returns to the chaotic motion. For this case the Melnikov theory gives a quite conservative estimation of chaos with \( F_0 \geq 0.25276 \). But we find that the chaos is happened after \( F_0 \geq 0.301 \).

(ii) For the value of \( F_0 = 0.2895 \) which is near to the starting value of chaos \( F_0 = 0.301 \), it leads to a higher subharmonic motion through period doubling route. However, the fourth-order Runge-Kutta method with a time stepsize \( h = 0.005 \) gives an incorrect 1/4 subharmonic motion as shown in Fig. 16(a), while the GPS2 can reveal a higher subharmonic motion in Fig. 16(b). The steady state motions are shown in a time interval of \( t \in [500, 2000] \).

After \( F_0 = 0.2676 \) the Duffing equation comes to a period-doubling range until the entrance to a chaotic state at \( F_0 = 0.301 \). Then we come to a narrow range of 1/3 subharmonic motion in the range of \( F_0 \in [0.356, 0.359) \), and then a 1/5 subharmonic motion in the range of \( F_0 \in [0.359, 0.38) \). In summary, we have found two ranges of \( F_0 \in [0.301, 0.356) \) and \( F_0 > 0.38 \) for the chaotic motions of the nonlinear Duffing oscillator.

7 Conclusions

In this paper the nonlinear differential equations system was converted into an augmented quasi-linear dynamical system in the Minkowski space, with the coefficient matrix being a Lie-form with \( B \in so(n,1) \). Based on the Lie-symmetry of the underlying new system, we have derived two types closed-form Lie-group solutions, of which the numerical scheme to preserve the Lie-group properties was developed. A signum function of sign\( (\|f\|^2\|x\|^2 - 2(f \cdot x)^2) \) was introduced. Then we have announced a very important concept of the first set of dis-connectivity in the hyper-plane \((x,f)\), whose variation with respect to the parameter of nonlinear dynamical systems as shown by the Duffing equation and the ship rolling equation reveals special structures. In view of Eqs. (43) and (59), the set of \( \mathcal{A}_1^- := \{ \text{sign}(\cos \theta) = +1 \text{ and sign}(\cos 2\theta) = +1 \} \) plays a dominant role for disclosing the complexity of nonlinear dynamical systems. The barcode can be used to detect the appearance of chaotic motion, of which the peak in the curve of the percentage to stay in the set of \( \mathcal{A}_1^- \) is an important criterion to forecast the subharmonic motions and chaos. One of the major contribution of this paper was that the proposal of the bifurcation diagram obtained by plotting the percentage of the first set of dis-connectivity \( \mathcal{A}_1^- \) with respect to the amplitude of external loading,
which leads to a finer structure of the devil staircase for the quadratic oscillator, and
the cascade of subharmonic motions to chaotic motions for the Duffing oscillator.
Hence, one has a practical and convenient tool to assess the complex behavior of
nonlinear oscillators.

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