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# A New Equalization Method for Lithium-Ion Battery Packs Based on CUK Converter

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# ABSTRACT

Aiming at the traditional CUK equalizer can only perform energy equalization between adjacent batteries, if the two single batteries that need to be equalized are far away from each other, there will be the problem of longer energy transmission path and lower equalization efficiency, this paper optimizes the CUK equalizer and optimizes its peripheral selection circuit, which can support the equalization of single batteries at any two positions. The control strategy adopts the open-circuit voltage (OVC) of the battery and the state of charge (SOC) of the battery as the equalization variables, and selects the corresponding equalization variables according to the energy conditions of the two batteries that need to be equalized, and generates the adaptive equalization current with an adaptive PID controller in order to improve the equalization efficiency. Simulation modeling is performed in Matlab/Simulink 2021b, and the experimental results show that the optimized CUK equalizer in this paper improves the equalization time by 25.58% compared with the traditional CUK equalizer. In addition, compared with the mean value difference (MVD) method, the adaptive PID method reduces the equalization time by about 30% in the static and charging and discharging experimental environments, which verifies the superiority of this equalization scheme.

# **KEYWORDS**

battery balancing; CUK equalizer; adaptive PID; SOC

# 1 Introduction

With the rapid development of society, people's demand for energy is increasing, and all walks of life around the world are gradually transforming into low-carbon [1-5]. Lithium-ion batteries have a series of advantages such as high energy density, long cycle life, clean and pollution-free, and are used in electric vehicles, aerospace and other industries widely [6-9]. Since single-cell lithium-ion batteries can provide limited voltage, in order to meet the needs of use, single cells are usually connected in series into groups [10]. Due to the inconsistency of parameters such as the production process, use environment, internal resistance of the battery, and self-discharge rate of a single battery, as the use time of the battery pack will be inconsistent, reducing the service life of the battery pack, so the battery balancing technology came into being [11,12].

From the perspective of energy, the balancing method of batteries can be mainly divided into two categories, namely active equalization (energy transfer equalization) and passive equalization (energy dissipative equalization) [13]. Passive equalization involves connecting resistors in parallel at both ends



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of the battery, which dissipates excess energy as heat. However, this method is not energy-efficient and can cause the battery to overheat, shorten its lifespan, and even explode in extreme cases [14]. As a result, active equalization has become a popular research topic. Active equalization transfers energy using an energy storage element, reducing battery pack inconsistency [15].

The active equalization circuit mainly includes four structures: inductor, capacitor, transformer and converter. Inductive equalization utilizes an inductor to transfer energy to a single cell or battery pack. If the batteries to be equalized are far apart, the energy consumed in the equalization process will increase and the equalization time will also increase [16]. Wu et al. [17] utilized a capacitive equalization topology, allowing electrical energy to flow between batteries. Similar to inductive equalization, when the first and last batteries are distant, the equalization process necessitates sequential equalization of adjacent batteries, leading to an increase in the number of times the equalization switch operates and the amount of energy consumed and lost in the circuit. Qian et al. [18] employed a bi-directional forward converter topology, known for its short equalization time and high efficiency. However, the presence of leakage inductance in the transformer must be taken into account. Cao et al. [19] favored a flyback transformer topology, which necessitates more transformers and raises the overall cost when the number of battery packs is higher. Zhang et al. [20] proposed a Buck-Boost converter equalization circuit that enables equalization between non-adjacent batteries and enhances the equalization efficiency. Nonetheless, the circuit's numerous switching devices make it challenging to achieve accurate control. Dam et al. [21] proposed and improves the CUK equalization circuit to achieve equilibrium between adjacent cells.

In this paper, an improved CUK equalization circuit is proposed, which can achieve equalization between arbitrary single cells, shorten the equalization path, improve the equalization efficiency, and have high topology compared with the traditional CUK equalizer. In terms of control strategy, the control strategy combining voltage and SOC is adopted, and different control variables are used at different stages to improve accuracy and reduce control difficulty. In terms of control algorithm, the adaptive PID control algorithm is adopted to adjust the conduction and turn-off of the MOS tube in the CUK circuit in real time to adjust the equalization current and improve the equalization efficiency.

## 2 Balanced Topology Analysis

# 2.1 The Topology of the Equalization Circuit

The classic CUK equalizer topology is illustrated in Fig. 1, consisting of n cells and n-1 CUK circuits, allowing energy transfer only between adjacent cells. When the batteries requiring equalization are distant from each other, energy must be transferred sequentially through neighboring batteries, resulting in lengthy equalization time and low efficiency.

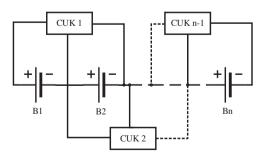


Figure 1: Traditional CUK equalizer

In this paper, an optimized CUK equalization circuit is designed consisting of a battery module, a selection circuit (discharge side and charging side), and a CUK circuit, as shown in Fig. 2. The battery module consists of *n* cells connected in series. The selection circuit is a switch array of double-layer switches, with switch  $K_{1i}$  on the discharge side and switch  $K_{2i}$  on the charge side connected to the positive and negative terminals of the battery. The main loop of the equalization topology is the CUK circuit, which is formed by inductors  $L_1$  and  $L_2$ , capacitor  $C_1$ , diodes  $D_1$  and  $D_2$ , and MOS tube  $Q_1$ .

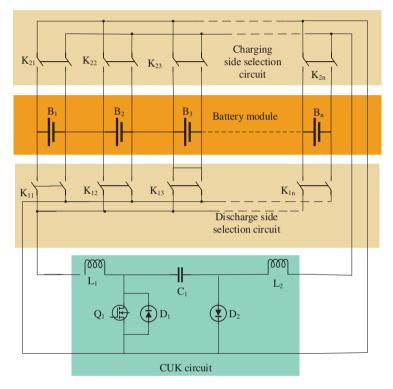


Figure 2: Improved CUK equalization circuit

# 2.2 Topology and Operational Analysis

The selection circuits of the balanced topology are respectively charging side and discharge side, which are connected to the positive and negative terminals of the battery and the CUK circuit respectively, the switch  $K_{1i}$  on the conduction discharge side with high battery energy, the switch  $K_{2i}$  on the conduction charging side with low energy, only one switch conducts at a time on the charging and discharging sides.

Taking battery  $B_1$  with the highest energy and battery Bn energy as examples, the circuit operating conditions are analyzed. When the energy of battery  $B_1$  is the highest, the switch  $K_{11}$  on the discharge side corresponding to battery  $B_1$  is turned on, so that it is connected to one end of the CUK circuit, the battery Bn energy is the lowest, and the switch  $K_{2n}$  of the charging side of the battery Bn is turned on and connected to the other end of the CUK circuit, and the result is shown in Fig. 3.

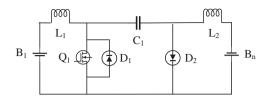
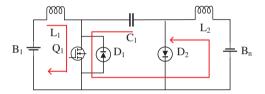


Figure 3: Selecting the completed equalization circuit

The turn-on and turn-off of switch  $Q_1$  is controlled by pulse width modulation (PWM). When  $Q_1$  is on, battery  $B_1$ , inductor  $L_1$  and switch  $Q_1$  form a loop; diode D2 cut off by reverse voltage, and capacitor  $C_1$ , inductor  $L_1$ , battery  $B_n$ , and switch  $Q_1$  form a loop, as shown in Fig. 4. At this time, battery  $B_1$  is discharged, inductor  $L_1$  is energy storage, and the discharge current is  $I_1$ ; Capacitor  $C_1$ , switch  $Q_1$ , battery  $B_n$  and inductor  $L_2$  form a loop, at this time, capacitor  $C_1$  charges battery  $B_n$ , inductor  $L_2$  stores energy, and the charging current is  $I_2$ .



**Figure 4:** The on-state of the switch  $Q_1$ 

When  $Q_1$  is turned off, diode  $D_1$  is cut off due to back voltage, and battery  $B_1$ , inductor  $L_1$ , capacitor  $C_1$  and diode  $D_2$  form a loop; The battery Bn, inductor  $L_2$  and diode  $D_2$  form a loop. This is shown in Fig. 5. At this time, battery  $B_1$  and inductor  $L_1$  charges capacitor  $C_1$  through  $I_1$ ; Inductor  $L_2$  charges the battery  $B_n$ , the current is  $I_2$ , and diode  $D_2$  acts as a current continuator.

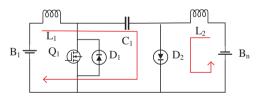


Figure 5: Switch Q<sub>1</sub> off state

According to the CUK equalizer circuit design standard, the amplitude of current fluctuations of inductors  $L_1$  and  $L_2$  should be less than 30% of the average current, see Eqs. (1), (2).

$$\Delta i_{L1} < 0.3I \tag{1}$$

$$\Delta i_{L2} < 0.3I \tag{2}$$

During the conduction of switching tube  $Q_1$ , it can be deduced from the power relationship that

$$\Delta i_{L1} = \frac{V_{\rm in}}{L_1} DT_S \tag{3}$$

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During the shutdown of the switch  $Q_1$ , it is available according to the power relationship.

$$\Delta i_{L2} = \frac{V_{\text{out}}}{L_2} (1 - D) T_s$$
(4)

Among them:  $V_{out}$  is the output voltage of the CUK equalizer, D is the duty cycle, and Ts is the switching period.

According to the characteristics of the CUK circuit, the input-output relationship can be obtained (5).

$$V_{\rm in} = \frac{D}{1 - D} V_{out} \tag{5}$$

Therefore, the value range of inductors  $L_1$  and  $L_2$  is

$$L_1 > \frac{V_{\rm in} DT_s}{0.3I} \tag{6}$$

$$L_2 > \frac{V_{\rm in} DT_{\rm s}}{0.3I} \tag{7}$$

The amplitude of the capacitor voltage fluctuation should be less than 5% of the average value, and the calculation formula of the intermediate storage capacitance  $C_1$  is as follows. During  $DT_s < t < T_s$  inductor  $L_1$ , battery  $B_1$ , and capacitor  $C_1$  form a loop, and this circuit charges capacitor  $C_1$ , depending on the power relationship:

$$\Delta V_{C1} = \frac{1}{C_1} I (1 - D) T_s < 0.05 V_{C1}$$
(8)
$$V_{in}$$
(9)

$$V_{C1} = \frac{r_{in}}{1 - D}$$
(9)

Therefore, the value range of capacitor  $C_1$  is

$$C_1 > \frac{I\left(1-D\right)^2 T_s}{0.05 V_{B1}} \tag{10}$$

The inductor average currents  $I_{L1}$ ,  $I_{L2}$  and capacitor voltage  $V_C$  are given by Eqs. (11)–(13).

$$I_{L1} = 0.5T_s \left( \frac{V_{B1}}{L_1} D^2 + \frac{V_C - V_{B1}}{L_1} \left( 1 - D \right)^2 \right)$$
(11)

$$I_{L2} = 0.5T_{S} \left( \frac{V_{Bn}}{L_{2}} \left( 1 - D \right)^{2} + \frac{V_{C} - V_{Bn}}{L_{2}} D^{2} \right)$$
(12)

$$V_c = \frac{V_{in}}{1 - D} \tag{13}$$

where  $V_C$  is the voltage of capacitor  $C_1$ , and  $V_{B1}$  and  $V_{Bn}$  are the voltages of the batteries connecting the two ends of the CUK circuit, respectively.

In summary, it can be seen that when its switching period, battery voltage, inductor-capacitor parameters and so on are determined, the equalization current is the size of the  $Q_1$  control signal is only related to the duty cycle D and period  $T_s$ .

#### **3 Balanced Control Strategy Design**

1

Equilibrium variables are the basis for the study of equilibrium control strategies, mainly battery open circuit voltage (OVC) and SOC [22]. Reference [23] was based on the battery open-circuit voltage as the starting or ending equilibrium control variable, although the voltage can be measured directly, but when the voltage is in the plateau period, the amplitude of the battery SOC change with voltage is not significant, and the real-time performance of the open-circuit voltage is also insufficient. SOC is the most characteristic of battery power, and it is also a commonly used indicator in equilibrium research. Jiaqiang et al. [24] compared and analyzed the voltage and SOC equalization under the same conditions, and the control effect of the latter is better under the premise that the accurate estimation results of the state of charge can be obtained. Although the SOC needs to be calculated, it can better reflect the current state of the battery and has a good equalization effect. This paper selects variables combining battery voltage and SOC, and uses different control variables at different stages to improve accuracy and reduce control difficulty. When the battery SOC is  $20\% \sim 80\%$ , SOC is selected as the equilibrium variable; when the battery SOC is  $0\% \sim 20\%$  and  $80\% \sim 100\%$ , the battery voltage is selected as the equilibrium variable.

The parameter setting of traditional PID control relies on the accurate control system model, the calculation process is complex, the test process needs to be manually tuned by trial method, relying on debugging experience, and the traditional PID controller is unchanged after the three parameters are set, and does not have adaptive characteristics. Adaptive PID uses online identification to identify changes in system parameters, and then uses the idea of pole configuration to calculate PID parameters to ensure optimal work. In this article, the adaptive PID controller can automatically adapt the equalization current to improve the equalization speed. Compared with fuzzy control, adaptive PID control has simple structure, good robustness and high reliability.

In this paper, the equalization control strategy is to use the adaptive PID control algorithm to adapt the duty cycle of PWM, so as to adjust the equalization current, when the difference between the voltage/SOC of the discharged battery and the rechargeable battery is large, the equalization current should be large, and vice versa.

As shown in Fig. 6, two types of equilibrium variables are considered, but the adaptive PID control is similar, with 2%  $\varepsilon$  and 10 mV  $\gamma$ . Taking the battery SOC as an example of the equilibrium variable, the battery SOC monitoring module will monitor the SOC in real time. When the battery's state of charge (SOC) is between 20% and 80%, the equalization circuit will connect the maximum battery SOC to the discharge side through the selection circuit, and the smallest battery SOC is connected to the charging end. The absolute value  $\Delta$ SOC of the difference between the average SOC<sub>ave</sub> of the SOCs of the two cells participating in the equalization and the SOC<sub>i</sub> of one of the cells participating in the equalization is used as the input of the adaptive PID, see Eqs. (14), (15), and its output can obtain the duty cycle of the PWM wave, thereby controlling the conduction and turn-off of the switch to achieve the effect of controlling the equalization current, with voltage as the equilibrium variable.

$$SOC_{ave} = \frac{1}{2} \left( SOC_i + SOC_j \right) \tag{14}$$

$$\Delta SOC = |SOC_i - SOC_{ave}| \tag{15}$$

In Eqs. (14), (15), SOC<sub>i</sub> and SOC<sub>i</sub> represent two cells participating in equalization.

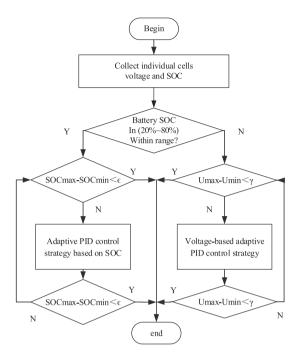


Figure 6: Balancing flow chart of lithium-ion battery pack based on CUK converter

## 4 Designing Simulation Platforms and Conducting Experimental Analyses

In this paper, the simulation experiment uses Matlab/Simulink2020b for simulation modeling, The lithium-ion battery parameters are 3.7 V/2.2 AH, with an initial SOC value for each individual cell. of the battery module are 78%, 74%, 70%, 66%, 62% and 58%, respectively. The equalization simulation model includes a battery module, a selection circuit, a CUK equalizer, a select circuit switch control module, and an adaptive PID-based current control module. Battery module, selection circuit, CUK equalizer are explained in Section 2.1; the selection circuit switch control module is based on the state of each battery in the battery module to achieve the switch on and off; the current control module, which is based on an adaptive PID, is implemented as a Matlab function. To validate the effectiveness of the proposed topology and algorithm in this paper, both the PID control module and the PWM output module are utilized. the following will be respectively the comparative experiment of balanced topology and adaptive PID control. The main parameters of the experimental circuit are shown in Table 1.

Table 1: Main component parameters of experimental circuit

Components	Parameter		
Inductors $L_1, L_2$ Capacitor $C_1$ Diodes $D_1, D_2$ MOS tube $Q_1$ Battery $B_1 \sim B_n$	220 uH 150 uF IN5822 IPD135N03L 10 Ah/3.2 V		

#### 4.1 Validation of Topology Performance

The study aims to confirm the effectiveness and superiority of the new CUK equalizer over the traditional CUK equalizer. A comparison experiment is designed to assess the equalization time needed for both topologies in stationary equalization. This will provide evidence of the new CUK equalizer's superiority.

Figs. 7 and 8 show the SOC curves of each cell in static equalization of the traditional CUK equalizer and the new CUK equalizer proposed in this paper, respectively.

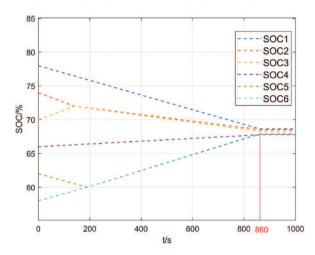


Figure 7: Traditional CUK static equilibrium

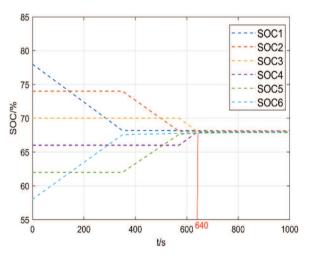


Figure 8: Novel CUK static equilibrium

Figs. 7 and 8 demonstrate that the traditional CUK equalizer takes approximately 860 s to complete equalization in a static state. In contrast, the new CUK equalizer proposed in this paper only takes 640 s to complete the same task, resulting in a 25.58% reduction in equalization time. These results confirm the effectiveness and superiority of the new CUK equalizer proposed in this study.

Compared with the traditional CUK equalizer, the topology in this paper uses double-layer switches as the selection circuit, despite the large number of switches, it is possible to reduce the number

of capacitive inductors, but the control is simple, the stability is strong, and the structural complexity of the topology will not increase exponentially when the number of batteries increases.

# 4.2 Adaptive PID Superiority Verification

In this paper, in order to validate the superiority of the adaptive PID control algorithm used, comparative experiments are carried out under three working conditions: stationing, charging and discharging, and the mean difference method is used for the comparative experiments. In this scenario, both the charge equalization test and the discharge equalization test employ constant current for charging and discharging static equalization.

Figs. 9 and 10 are based on the mean value difference (MVD). method based on the static equilibrium cell SOC curve and the static equilibrium cell SOC curve based on adaptive PID.

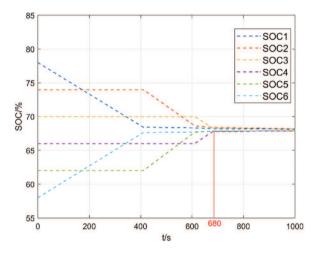


Figure 9: Stationary equilibrium based on mean difference method

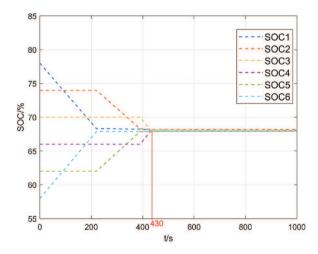


Figure 10: Static equalization based on adaptive PID

From Figs. 9 and 10, it can be observed that under the static equalization experimental conditions, it takes approximately 680 s to complete the equalization using the mean-value difference method,

whereas it only takes 430 s to complete the equalization using adaptive PID control, resulting in a 36.76% improvement in equalization speed. This verifies the superiority of adaptive PID control over the mean-value difference method in static equalization.

a. Charge equalization

Figs. 11 and 12 are the SOC curves of each cell of charge balancing based on the MVD method and the SOC curve of each cell of each cell based on adaptive PID, respectively.

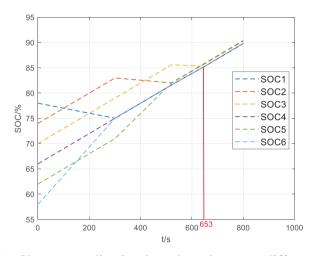


Figure 11: Charge equalization based on the mean difference method

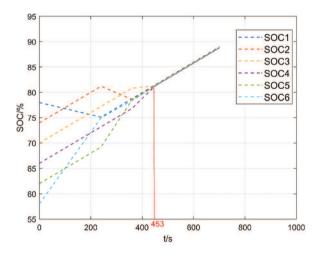


Figure 12: Charge equalization based on adaptive PID

From Figs. 11 and 12, it is evident that in the charging equalization experiment, it takes approximately 653 s to complete the equalization using the mean-value difference method, whereas it only takes 453 s to complete the equalization using the adaptive PID. The equalization speed is improved by about 30.62%, confirming the superiority of the adaptive PID control over the mean-value difference in the charging equalization.

## b. Discharge equalization

Figs. 13 and 14 are respectively the SOC curve of each cell with emission equalization based on the MVD method and the SOC curve of each cell with emission equalization based on adaptive PID.

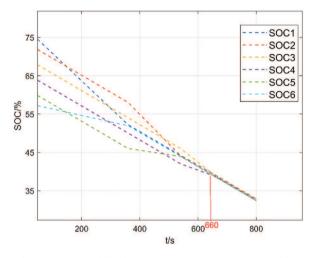


Figure 13: Discharge equilibrium based on the mean difference method

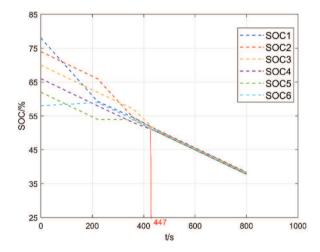


Figure 14: Discharge equalization based on adaptive PID

From Figs. 13 and 14, we can observe that in the discharge equalization process, it takes approximately 660 s to complete the equalization using the mean-value difference method, and about 447 s to complete the equalization with adaptive PID. This results in an improvement in equalization speed of about 28.78%. This experiment confirms the superiority of the adaptive PID control over the mean-value difference method in discharge equalization.

The SOC and SOC ranges of each single cell after the completion of the standing experiment, charging experiment and discharge experiment are shown in Table 2.

	Stationary experiments		Charging experiment		Discharge experiment	
	Mean difference	Adaptive PID	Mean difference	Adaptive PID	Mean difference	Adaptive PID
Battery1	68.97%	68.77%	89.87%	89.42%	33.61%	38.62%
Battery1	69.23%	68.83%	90.61%	89.56%	33.83%	38.22%
Battery1	68.88%	68.42%	90.43%	89.87%	32.94%	38.01%
Battery1	67.97%	68.31%	88.54%	88.78%	32.42%	37.83%
Battery1	67.09%	67.93%	89.88%	88.87%	31.62%	37.72%
Battery1	67.53%	67.51%	89.79%	88.74%	31.93%	37.42%
Range	2.14%	1.32%	2.07%	1.13%	2.21%	1.2%

 Table 2: SOC and SOC range of each single cell

In Table 2, it is evident that the polar deviation of each individual cell is smaller when using the adaptive PID control method compared to the mean-value difference method in static equalization, charging equalization, and discharging equalization experiments. Therefore, the equalization effect is better when based on adaptive PID control.

From Figs. 9–14 and the data analysis in Table 2, it is evident that the adaptive PID algorithm's effectiveness lies in its ability to dynamically adjust the equilibrium current and enhance the equilibrium speed in static, charging, and discharging experiments.

# 5 Conclusion

Aiming at some problems of the traditional CUK equalizer in the equalization process, this paper carries out some optimization of the CUK equalizer, which can realize the equalization between any two batteries, shorten the equalization path and improve the efficiency. In terms of equalization variable selection, battery voltage and battery SOC are used as composite equalization variables, and SOC is used as the equalization variable in the voltage plateau period, while battery voltage is used as the equalization variable in other periods. Adaptive PID control strategy is used to generate adaptive equalization current, shorten the equalization time and improve the equalization efficiency. The effectiveness of the optimized topology in this paper is experimentally verified.

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Availability of Data and Materials: The data that support the findings of this study are available from the corresponding author upon reasonable request.

**Conflicts of Interest:** The authors declare that they have no conflicts of interest to report regarding the present study.

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